Appendix I Data Sheets

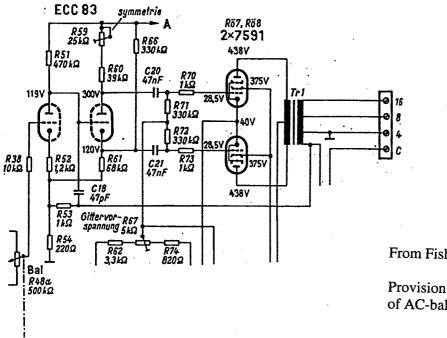
3 MQ' 22 MQ' U/k (k neg) - 50 V In Phasenumkehr. stufen max, 120 kΩ stages max. 120 kG 6 mA Ú/k (k pos) - 100 V 10 MQ **U**s1 only produced 20 kQ 0,2 W · In phase-splitter 550 V 200 650 V **200 V** 1 selbstanlaufend Maximum Ratings (Qa > 0.2 W) * (Q4 < 0,2 W) I • Grenzdaten by Rg1 Uakalt -Ugg kalt" Rfk 2 2 ð بع 1 MD 0,9 mA 2,2 kΩ 46 Veff 680 kΩ 7 1 MΩ Characteristics and Typical Operation 220 KD Rundfunk- und Fernsehverstärkerröhren, Gteig 250 V 180 The harmonic distorion is approx, proportional to the output voltage ist der Ausgangsspannung Ub - 200 250 200 ¹ Bei k - 5%, der Klirrfaktor 8 -88 2,2 . **0,**8 36 12 Aligemeine Daten | Kenn- und Betriebsdaten . **1**0 8 g 0,39 21 112 - 2,0 mAV etwa proportional -22 - 30 mA - 0,6 mA - 2,5 MQ **Typical Operation** Casi < 0.05 pF Betriebadaten - 140 V 20 - 260 V 330 Ua~ - 40 Characteristics 1 06 8 NF-Verstårker - 0,39 5 AF Amplifler 8 Kenndaten : ì I P 2281 2 RI n, 22 U.S. 5 `***** Ra 1. 2 2 C₂₁ < 0,0025 pF 4.0 pF Cause - 5.5 pF Capacitances Kapazitāten General Data UJ- 6.3 V - 0,2 A Indirekt ; Ceing -Heizung Heating Indirect 1 als NF-Verstärker Kolben Nr. 6 Verwendung AF Amplifler Bulb No. 6 AF Pentode NF-Pentode EF 86 Noval $(U_{g1} \sim - 0)$. 4 W $(U_{E1}\sim>0)$ 107 0,5 MQ** 0,7 MΩ* 150 mA 20 KΩ 8 W 1<u>8</u> < 0,3 MΩ* S3 mA 1 MΩ - 25 W) - 27,5 W ۶00 V 425 V 2 W 20 kΩ 100 < Uakalt - 2000 V Maximum Ratings 300 < 300 V. 100 < Grenzdaten Maximum Ratings 2008 12 W KI, A und AB Fixed grid bias Qa (Uz1~-0) Q4 (U2.~ >0) Grenzdaten orröhren Ugakalı – • U_{fi} fest ı •• KI, B - n⁻¹ Q.s 5 с**г** ్లి 0 5 ş ، ð 11 mAV 8,7 Veff - 11,3 mA/V 4,3 Veft 49,5 mA ren, Gleichi 0,5 V_{eff} 100 mA 14,9 mA Characteristics, and Typical Operation ã Characteristics and Typical Operation >0 15 kQ 2,0 KQ 10 % α 265 V ğ 11 W - 5,5 mA > -13,5 V R₁ = 40 kQ 250 V Ŧ #22£1= 19,5 Allgemeine Daten | Kenn- und Betriebsdaten 23 Kenn- und Betrlebsdaten -7,3 135 5,2 ទ 22 22 23 U21~N~(- 50 mW) Betriebsdaten **Typical Operation** ₽ 0,65 Ŧ **Typical Operation** c -145 R 0'6 8 с С Betrlebsdaten 2 **6**,3 202 22 U₂₃ - 250 V - 48 mA Uz - 250 V Rundlunk- und Fernsch Characteristics U_{g1} = -7.3 V ₽ Kenndaten . Elntakt A ł Eintakt A Class A 1 **~**"an 1315, Class A 2 ~**1**3∩ 5 ടീ I'a 5 ž 2 . 4 Aligemeine Daten Cause - 7,2 pF < 1,0 pF < 1.0 pF - 11 pF Ceing - 15,5 pF 6,5 pF - 10,8 pF 0,5 pF C_{**f**1} < 0.25 pF Capacitances General Data General Data - 6,3 V - 0,78 A Kapazitäten Capacitances *lf* = 1.5 A Kapazităten Ur- 63 V Indirect . Heating ٠. Indirect Helzung Helzung Heating Indirekt Cat Indirekt CFC CFC C_{df1} °, 5 5 ບັ for Power Amplifier für Kraftverstärker Power Pentode Power Pentode Kolben Nr. 23 Verwendung Bulb No. 23 Endpentode Endpentode いっこう ちょうちょうなをを Fortsetzung EL 84 6 BQ 5 EL 34 6 CA 7 Oktal ,

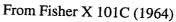
ECC 83	Aligemeine Daten General Data	Kenn- und Betriebsdaten Characteristics and Typical Operation	Grenzdaten Maximum Ratings		Allgemeine Daten General Data	Kenn- und Betriebsdaten Characteristics and Typical Operation	Maximum Ratings
IZ AA I NF-Doppeltriode		Kenndaten	je System	HF-Doppeltriode	Helzung	Kenndaten Chandaten	je System per section
Verwendung als				Verwendung als	Heating		TEO V
NF-Verstårker und	Ir = 0.3 A	-2,0 V # =			U/C 6,3 V	U_ = 250 V	U = 300 V
	oder	$I_d = 1,2 \text{ mA} R_i = 62,5 \text{ kg}$	I,		r or a la l	21 - 10 mA	ļ
AF Twin Trinde	Uf - 12,6 V		I	.`	Ur = 12,6 V	, 1	1
	If = 0,15 A	Betriebsdaten	ï	RF Twin Triode	Ir - 0,15 A	8	۲ ۱
AF Amplitler	Indirekt	Typical Operation		VHF Oscillator	Indirekt	R ₁ = 11 kg	1
Phase-splitter	Indirect	NF-Verstärker		VHF Mixer	Indirect	2 	
	Kanazitātan	AF Amplifier	f	RF Amplifler			K/k = 50 KK
1000	Capacitances	200 200			•		
	Carkr = 0,33 pF		- 22 MO		Kapazitäten	Ckuf -	2,5 pF
		- 680 680 680	U., only produced		Capacitances	- 1,9 pF CkII/(#II+/)	
		- 3,3 2,7 2,2	by R_1 - 22 MQ	316 UU 31/21	١.	Calall < 0.4 pF	4,7 pF
9 Ip					1	Calati v	
Noval	Calsi Calleli	1 - 0,30 0,48 0,55 0,55 min	stufen max. 150 kg		Cals1 = 1,6 pr		5 PF
V-H N 6		95	In phase-solitter		ļ		A DF
Kolben Nr. o Bulb No. 6	ver∫ = vert∫ < 150 mpF	k = 4,6 3,4 2,6	stages max. 150 kG	Kolben Nr. 6 Bulb No. 6	Ckr/(g1+))-4,8 pF	Cellell 0,20 pF	
ECC 85	Allgemeine Daten	Kenn- und Betriebsdaten	Grenzdaten	ECC 82	Aligemeine Daten	Kenn- und Betrlebsdaten	Grenzdaten
6 AD 8	General Data	Characteristics and Typical Operation	Maximum Ratinos	19 411 7	General Data	Characteristics and Typical Operation	Maximum Ratings
	Helzung	Kenndaten	je System	NF-Doppeltriode	Heizung / Heating	Kenndaten / Characteristics	je System
HF-Doppeltriode	Heating	Characteristics	per section	Verwendung als		Je System / per section	section
Verwendung als	U/- 6,3 V	U _e = 250 V S = 6.0 mA/V	U. Lat - 550 V	NF-Verstärker.	If - 0,3 A oder	Ą.	1
HF-Verstärker und	Ir - 0,435 A	2,2 V μ		Phasenumkehrröhre.	UJ - 12,6 V	U ₂ 8.5 V	Q 4 - 2,75 W
selbstschwingende	Indirekt	- 10,0 mA	1	SynchronIsations-	<i>If</i> - 0,15 A	Ia – 10,5 mA	ł
Mischröhre	Indirect		+0,	Trennröhre. Multivibrator	Indirekt / Indirect	S = 2,2 mA/V	1
•		Betrlebsdaten		und Sperrschwinger	Kapazitäten	L	N I
RF Twin Triode		Typical Operation	<i>Ik</i> = 15 mA	AE Turis Triado	Capacitances		$K_{g} = 1 MM$
RF Amnlifter	-	HF-Verstärker (System I)	ŗ		Ceingil - 1,8 pr	Betriebsdaten / Iypical Uperation NE-Voreiston / AE Amolishon	Brt = 20 k0**
Self-Excited Mixer	Kapazitäten	Amplifier (Section)	Ļ		Causell = 15 nF		pulsdaue
	Capacitances		I		Court < 135 mpF	100 100 100 100 100 100 100 100 100 100	einer Periode,
	C 21/(k1+f+s)		x/x = 50 kg		v		f max = 2 ms
-	I.				v	R, - 330 330 330 3	Pulse time - 10%
		<i>U</i> . – – – 2 V			ł	ו לי	per cycle,
•	Cark1 = 0,18 pF	ı			ĩ	2,2 2,2	f max = 2 ms
	$G_{II}^{(k_{II}++s)}$, 1		Noval	1	1,30 1,63 1	** In Phasenumkehr-
	Carign - 1.5 pF	1		Kolhan Nr. 6		11 - 17 05 30 41 V -	stufen max. 150 kg
	Callkii = 0,18 pF				Carteri < 110 mpF	k = 5,6 5,8 5,9	stages max. 150 kg
				•			

Appendix II

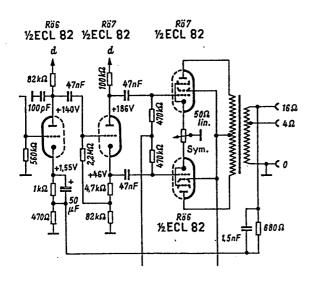
parts of diagrams The Marantz 8B The Williamson

The Quad II



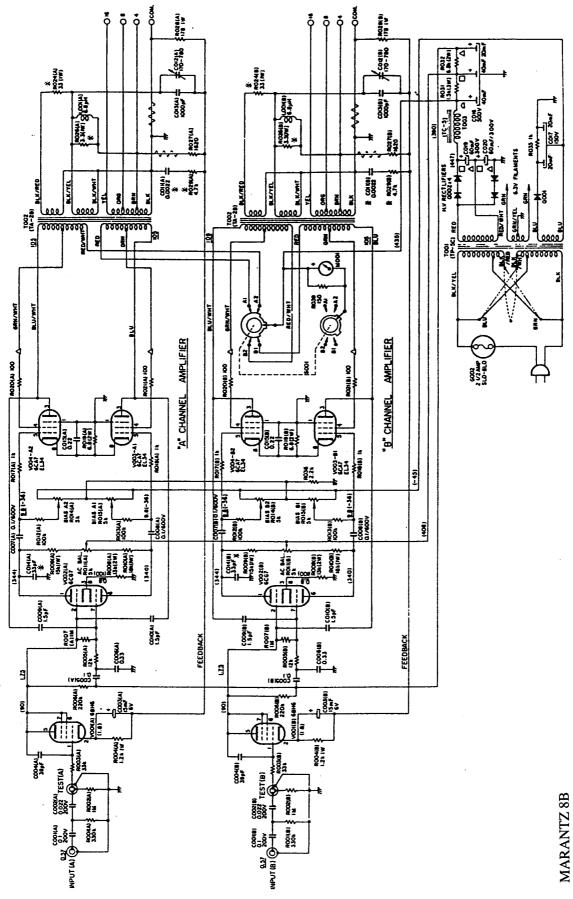


Provision for adjustment of AC-balance.

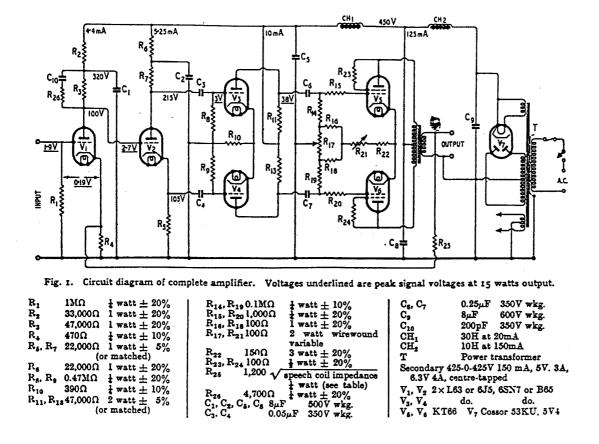


TELEWATT VS56 (1963)

Different load resistors in anode and cathode of the phase-splitter to compensate for difference in output resistances.

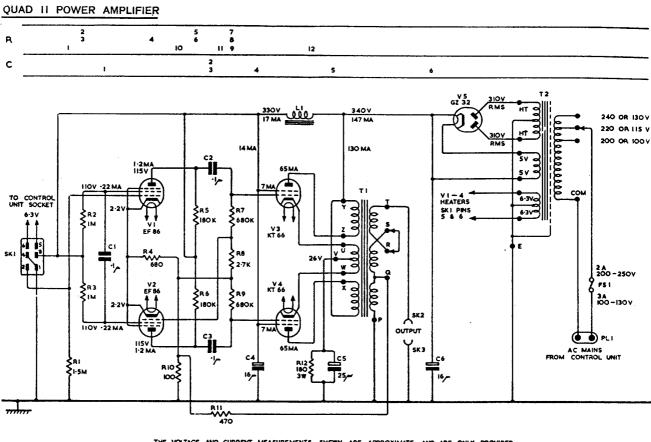


scheme as the amplifier described in this paper. in order to compensate for imperfections in the The global feedback system is very elaborated A very respected amplifier following a similar output transformer.



The Williamson Amplifier

D.T.N. Williamson was employed at Osram-Marconi, the company who developed and manufactured the K66 beam tetrode. His famous design was published in 1947 and it was a major step forward in terms of quality. Up to that time phase-splitting was normally performed by a transformer with a center-tapped secondary. The inclusion of two transformers in the signal path, made global feedback virtually impossible. The Williamson was able to deliver 15 Watts with a THD at 0.1% (at 1000Hz). It became a reference standard for more than a decade.



DRG 11175. ISSUE 1. THE VOLTAGE AND CURRENT MEASUREMENTS SHEWN ARE APPROXIMATE, AND ARE ONLY PROVIDED AS A GUIDE, ALLOWANCE SHOULD BE MADE FOR THE LOADING EFFECTS OF A VOLTMETER.

Shown here is the famous Quad II. It was a very well-made amplifier, and many of them are still in use after more than 30 years of service.

At first sight the diagram looks confusing, but after a closer look it becomes very simple.

The upper EF86 is quite a normal input stage. Signal for the lower EF86 is taken from the output of the upper via a voltage divider consisting of the 680 kw grid resistor of the upper KT66 and a 2.7 kw resistor.

The EF86^s both invert the signal so the driving voltages for the output stage are in opposite phase as required.

The voltage divider compensates for the gain of the lower EF86, where a gain of 680:2,7 » 250 apparently is expected. The actual gain is however much lower, but we find the explanation in the global feedback loop, where feedback is normal NFB to the upper EF86, but due to the inversion it is positive feedback to the lower EF86! This accounts for the high gain of that valve.

This combined input, phase splitter and driver stage is elegant, but the fact that the driving voltage to the lower EF86 has been exposed to HF cut-off and phase shift introduced by the loading of the output stage of the upper EF86 should not be overlooked.

It is unusual to tie the screen grids of the two valves together signal-wise instead of keeping them signal-free by connecting capacitors to ground from each grid, but this arrangement will keep the grids free from signal if they carry the same signal in opposite phase. If not, C_1 will help to restore AC balance from the two EF86^s.

Another unusual feature is the local feedback of the output stage by the cathode coils of the output transformer. It could be thought that this negative current feedback would raise output resistance of the entire stage. Because of the close coupling of all coils quite the opposite is the case. If output voltage decreases because of external load, the feedback voltage from the cathode coil will decrease too, which helps to restore the output signal. This local feedback of the output stage is a brilliant idea (but it raises demands for voltage swing on the grids!) and I should wish that such transformers were available. The ratio between anode – and cathode coils is approximately 10:1.

Despite its heavy weight and big transformers the Quad II was only rated as a 12W amplifier – and despite the measured performance is average, the sound is very good.

Because KT66 is hard to obtain and very expensive, it is often replaced by EL34 when Quad^s are restored. EL34^s works quite well, but remember to connect the 3rd grid to cathode as this is not done internally in EL34 (Pin 1 strapped to 8). Apart from that, the socket connections are the same. The 180w cathode resistor should be replaced by a 240w resistor (two 470w 5W resistors in parallel). As there is no provision for adjusting DC balance (and no self-balancing!) the output valves should preferably be matched. A 220w resistor in the supply line to the screen grids is recommended.

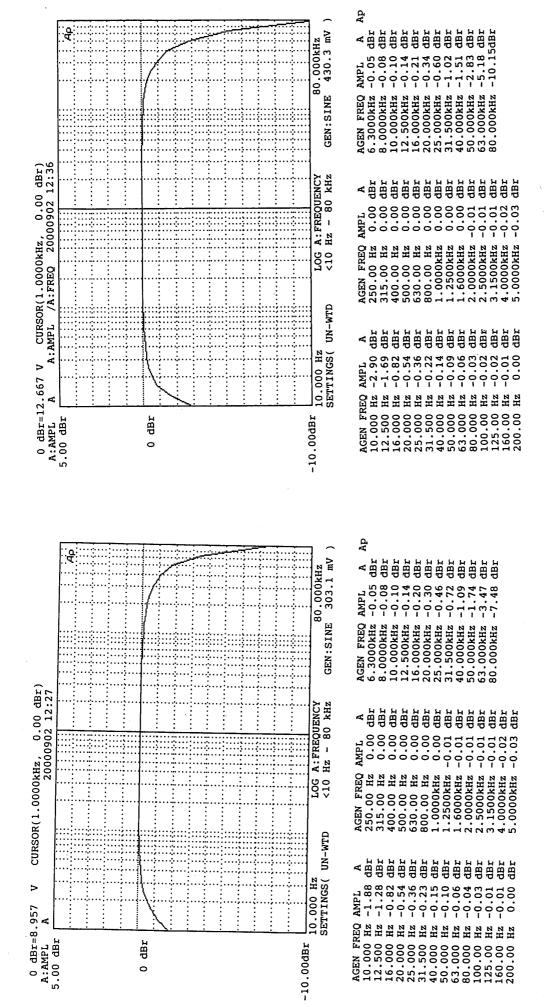
The GZ32 rectifier is often replaced by solid state diodes (1N4007), which also allows for a higher value of the surge capacitor C6, where 100m/450V would be a considerably improvement. Remember that this is not allowable with the GZ32 because it can't cope with the peak currents of a 100mF capacitor.

You cannot change the amount of feedback without changing the ratio of the voltage divider in the grid resistor of the upper KT66, and this is not advisable.

Despite reservations, I have always been fascinated by the circuit. Although it is easy to understand how it works, it has always been an enigma to me how it was conceived!

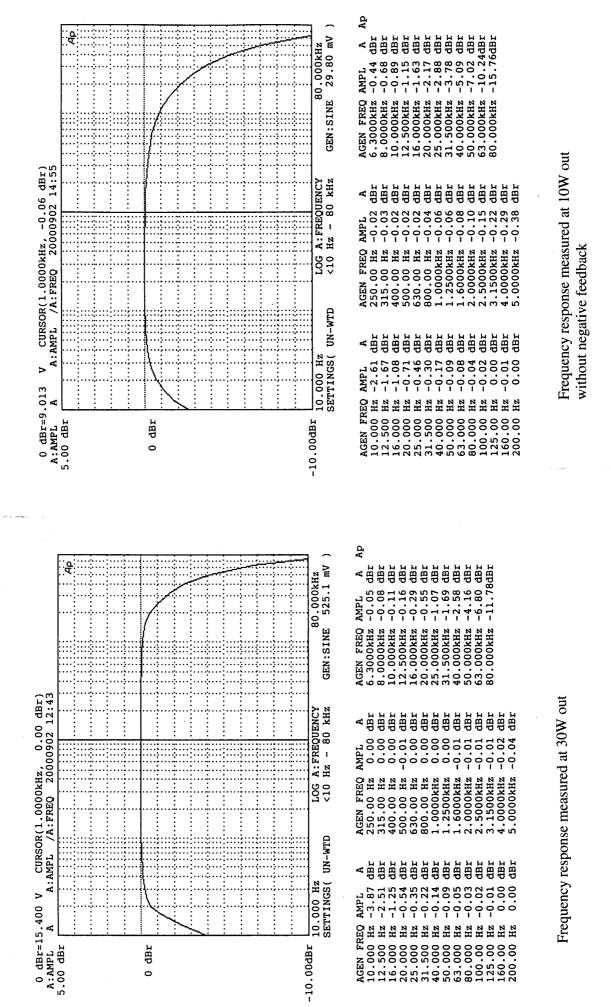
Appendix III

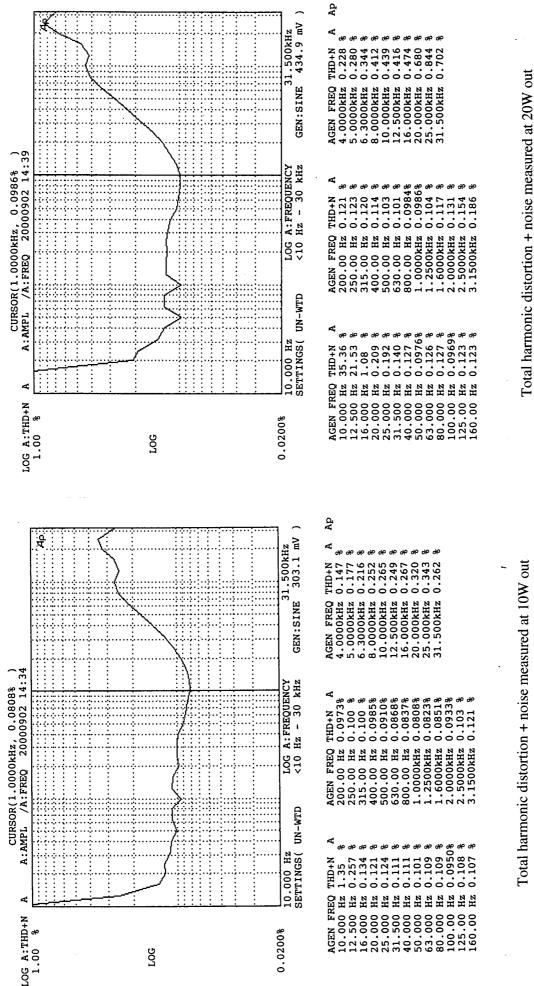
performance curves



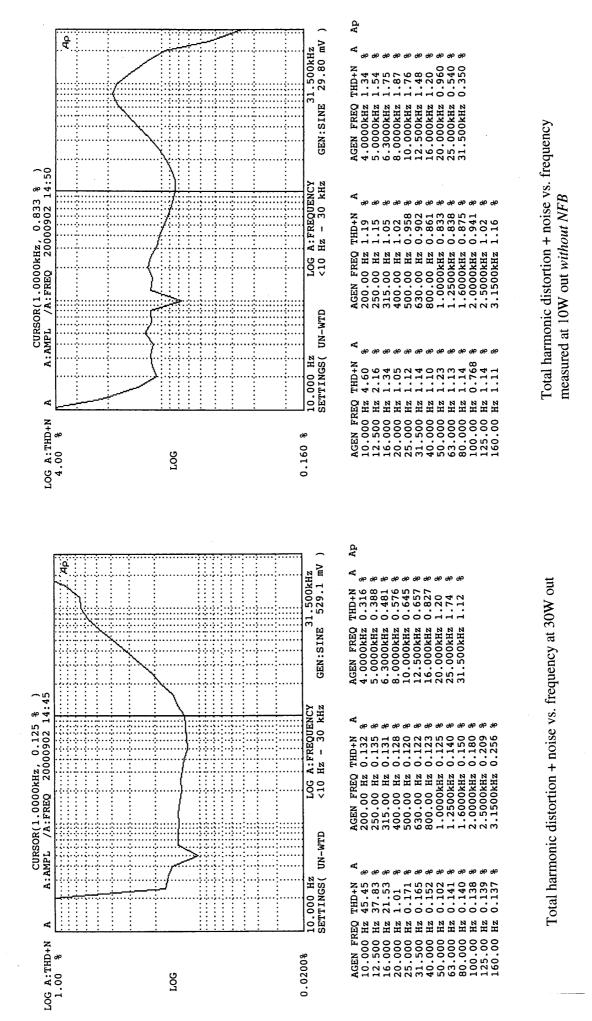
Frequency response measures at 20W out

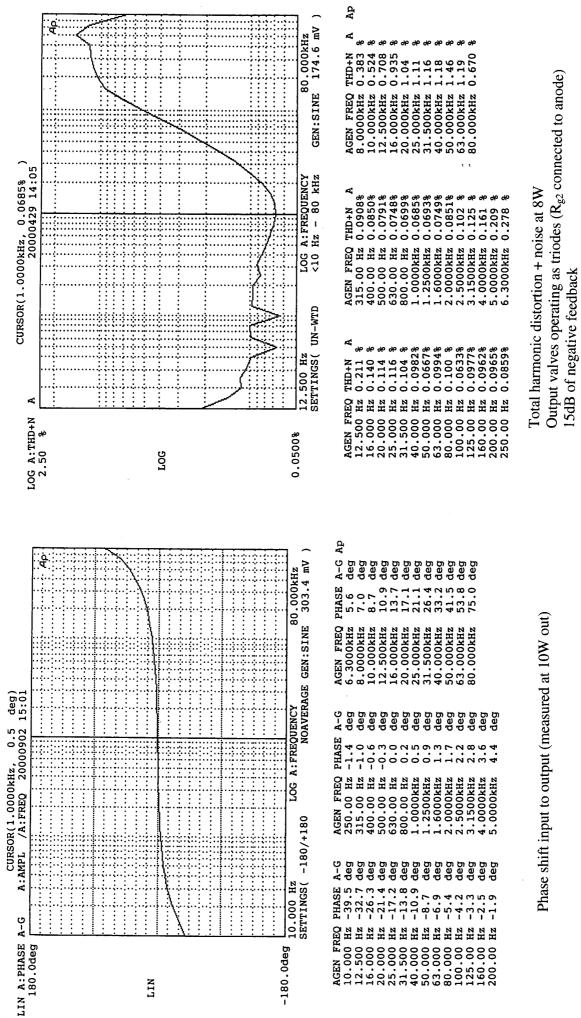
Frequency response measured at 10W out

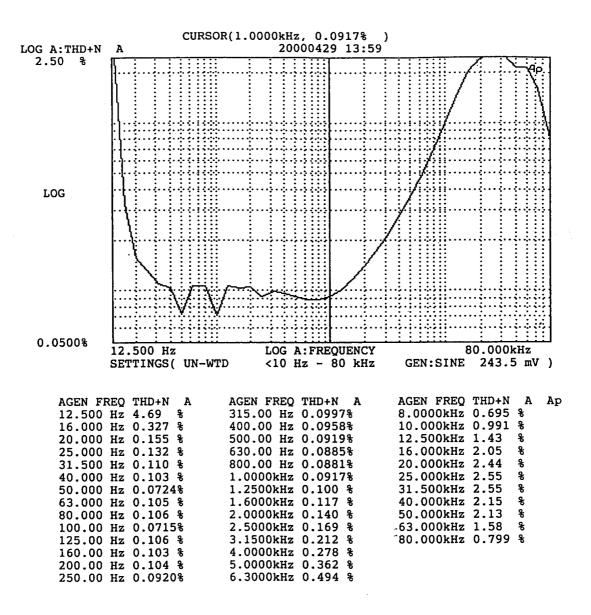




Total harmonic distortion + noise measured at 10W out

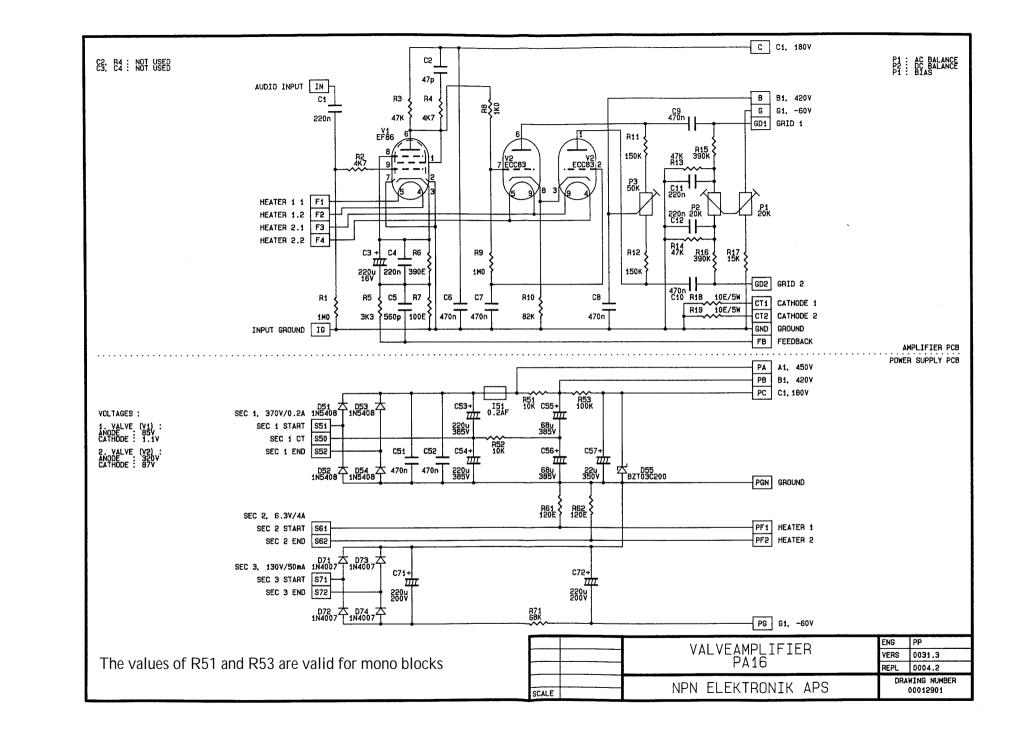


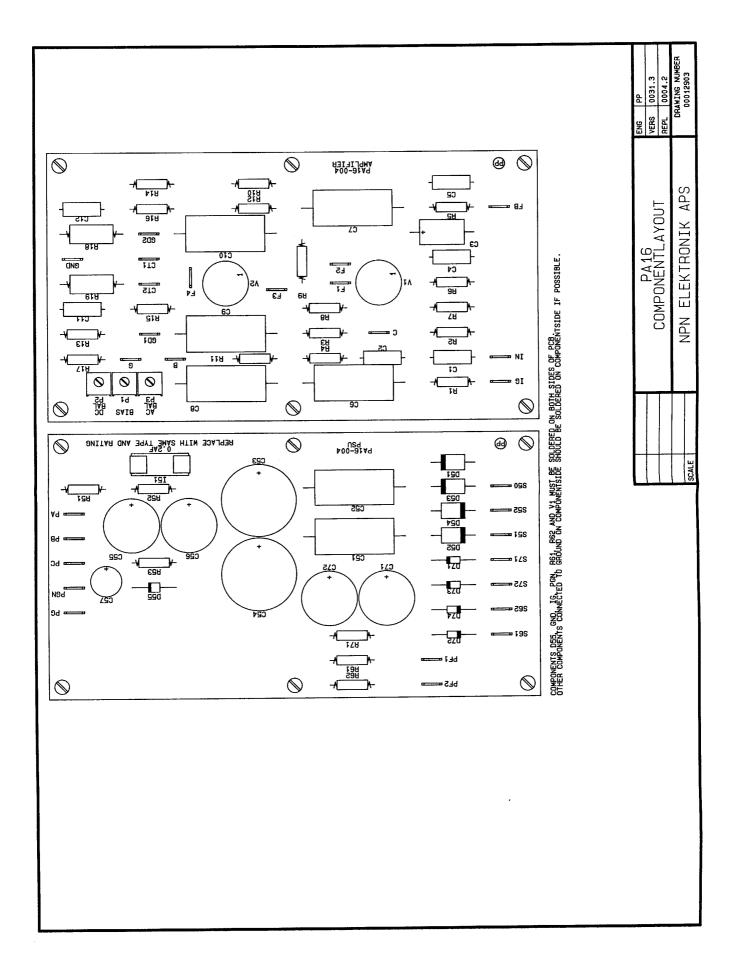




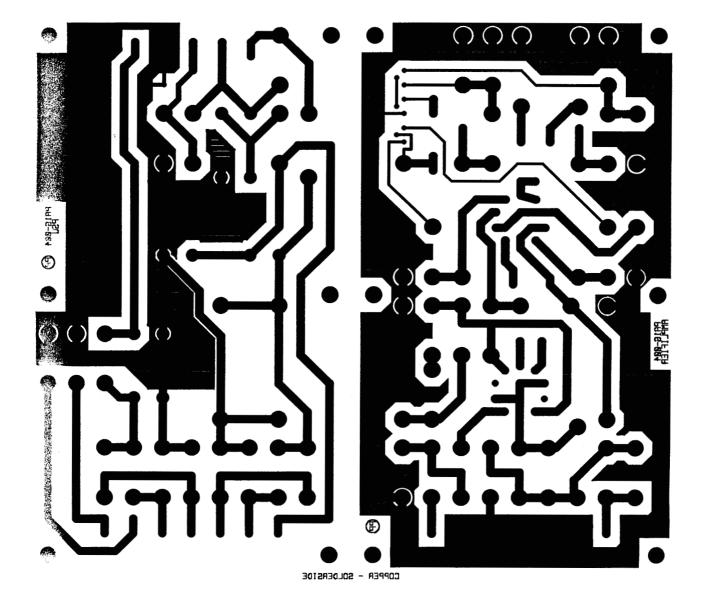
Total harmonic distortion + noise at 16W Output valves operating as triodes (R_{g2} connected to anode) 15dB of negative feedback

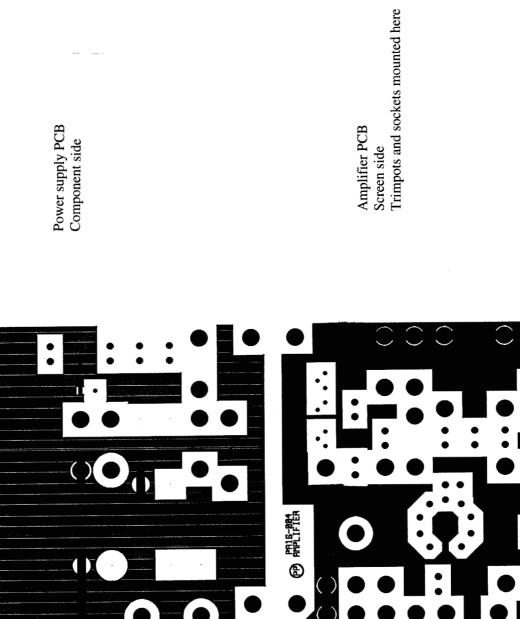
Appendix IV PCB's and layout





Power supply PCB Track side Amplifier PCB Track side resistors and capacitors mounted here





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